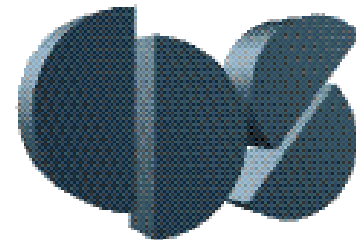




# CDS 110/ChE 105: Lecture 7-1

## Small Signal (Freq Domain) Analysis



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**13 May 2024**

### **Goals:**

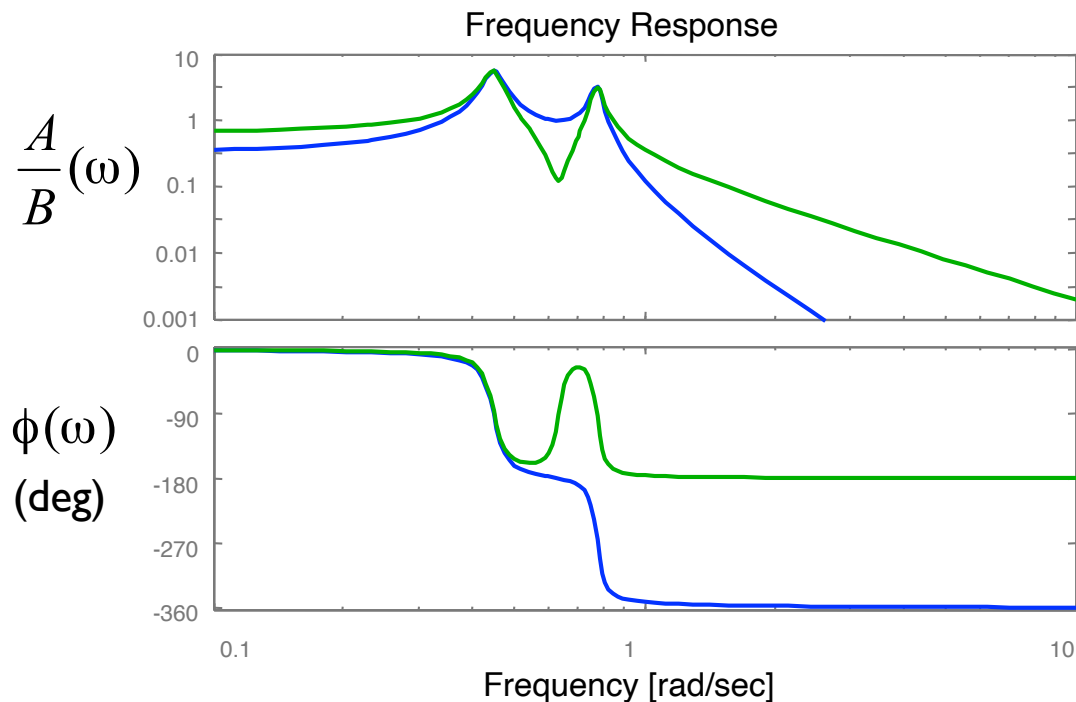
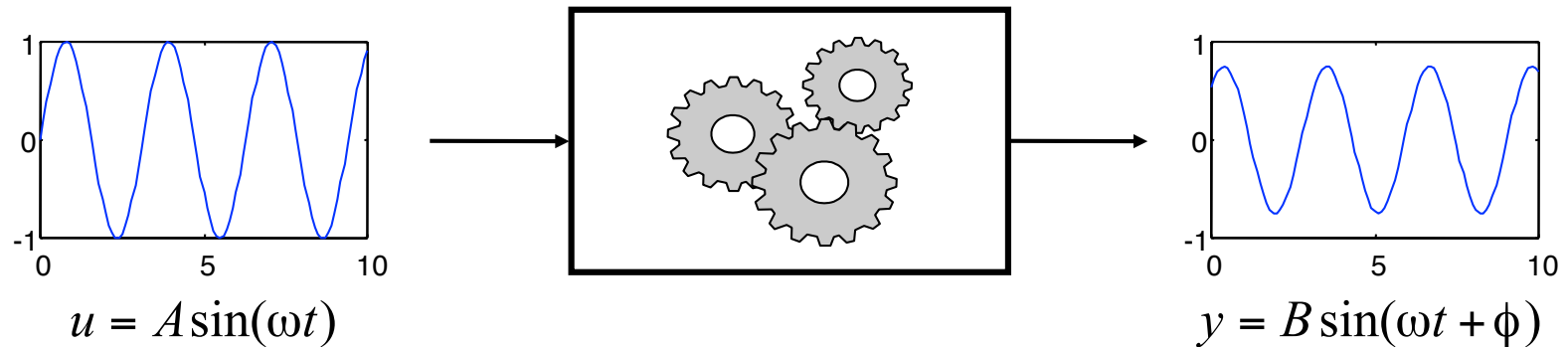
- Define the input/output transfer function of a linear system
- Describe how to use Bode plots to understand frequency response
- Understand relationships between frequency response, transfer function, and state-space model

### **Reading:**

- Åström and Murray, *Feedback Systems*, Sec 9.1-9.4

# Frequency Domain Modeling

**Defn.** The frequency response of a linear system is the relationship between the gain and phase of a sinusoidal input and the corresponding steady state (sinusoidal) output.



## Bode plot (1940; Hendrik Bode)

- Plot gain and phase vs input frequency
- Gain is plotting using log-log plot
- Phase is plotting with log-linear plot
- Can read off the system response to a sinusoid – in the lab or in simulations
- Linearity  $\Rightarrow$  can construct response to any input (via Fourier decomposition)
- Key idea: do all computations in terms of gain and phase (frequency domain)

# Transfer Function and Frequency Response

## Exponential response of a linear state space system

$$y(t) = \underbrace{C e^{At} \left( x(0) - (sI - A)^{-1} B \right)}_{\text{transient}} + \underbrace{\left( C(sI - A)^{-1} B + D \right) e^{st}}_{\text{steady state}}$$

## Transfer function

- Steady state response is proportional to exponential input => look at input/output ratio
- $G(s) = C(sI - A)^{-1} B + D$  is the *transfer function* between input and output

## Frequency response

$$u(t) = A \sin \omega t = \frac{A}{2i} (e^{i\omega t} - e^{-i\omega t})$$

$$y_{ss}(t) = \frac{A}{2i} \left( G(i\omega) e^{i\omega t} - G(-i\omega) e^{-i\omega t} \right)$$

$$= A \cdot |G(i\omega)| \sin(\omega t + \arg G(i\omega))$$

gain

phase

## Common transfer functions

$$\dot{y} = u \quad \frac{1}{s}$$

$$y = \dot{u} \quad s$$

$$\dot{y} + ay = u \quad \frac{1}{s+a}$$

$$\ddot{y} = u \quad \frac{1}{s^2}$$

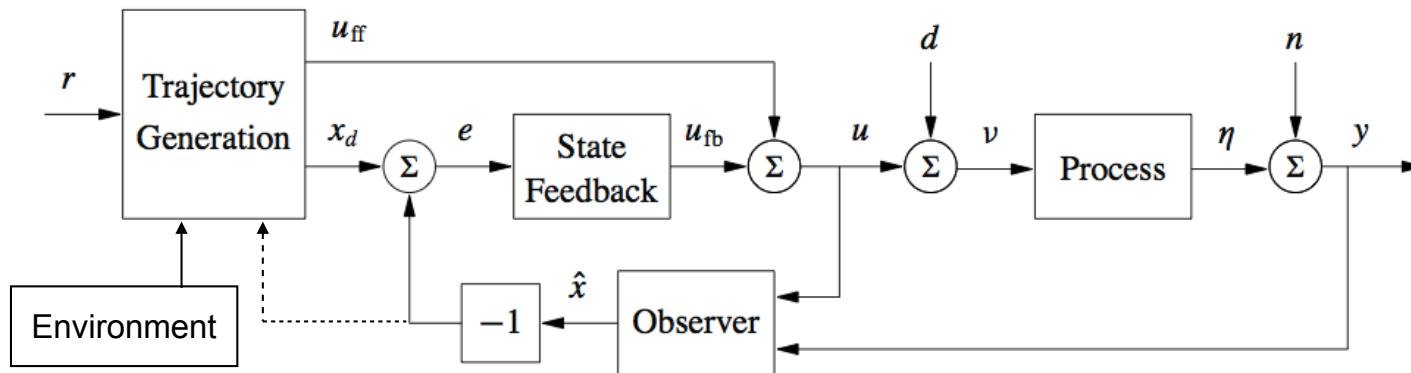
$$\ddot{y} + 2\zeta\omega_n\dot{y} + \omega_n^2 y = u \quad \frac{1}{s^2 + 2\zeta\omega_n s + \omega_n^2}$$

$$y = k_p u + k_d \dot{u} + k_i \int u \quad k_p + k_d s + \frac{k_i}{s}$$

$$y(t) = u(t - \tau) \quad e^{-\tau s}$$

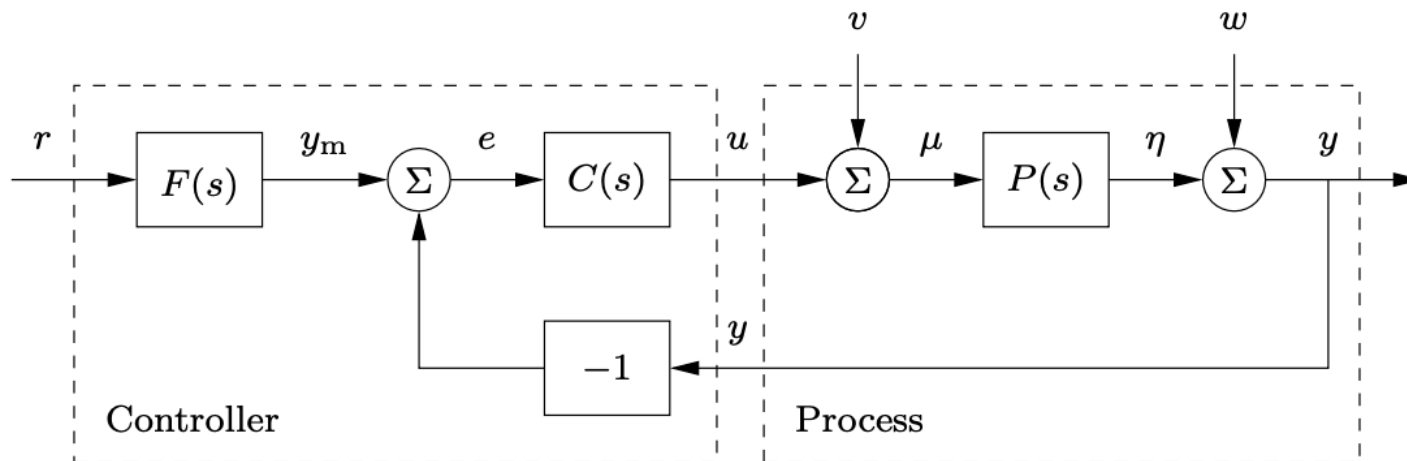
# Small Signal Analysis Using Transfer functions

## State feedback design pattern



- Allows (optimization-based) high performance trajectory generation and tracking
- Can be difficult to understand stability and robustness except via simulation...

**Small signal approach:** assume signals are small, analyze linear systems in frequency domain



- Focus on operation near equilibrium or along a trajectory
- Replace (nonlinear) input/output maps with transfer functions
- Enables robustness analysis via frequency bounds (next week)

- Early control theory *started* with this design pattern, driven by early advances in electronics
  - Use exponential response (Laplace transforms) to model dynamics and control
  - Use “loop shaping” (frequency domain design) for design: performance vs robustness

# Transfer Function Properties

**Def.** The transfer function for a linear system  $\Sigma=(A,B,C,D)$  is given by

$$G(s) = C(sI - A)^{-1}B + D \quad s \in \mathbb{C} \quad G(s) = \frac{n(s)}{d(s)} \quad d(s) = \det(sI - A)$$

**Thm.** The transfer function  $G(s)$  corresponding to  $\Sigma = (A,B,C,D)$  has the following properties:

- $G(s)$  is a ratio of polynomials  $n(s)/d(s)$  where  $d(s)$  is the characteristic equation for the matrix  $A$  and  $n(s)$  has order less than or equal to  $d(s)$ .
- The steady state frequency response of  $\Sigma$  has gain  $|G(j\omega)|$  and phase  $\arg G(j\omega)$ :

$$u = A \sin(\omega t)$$

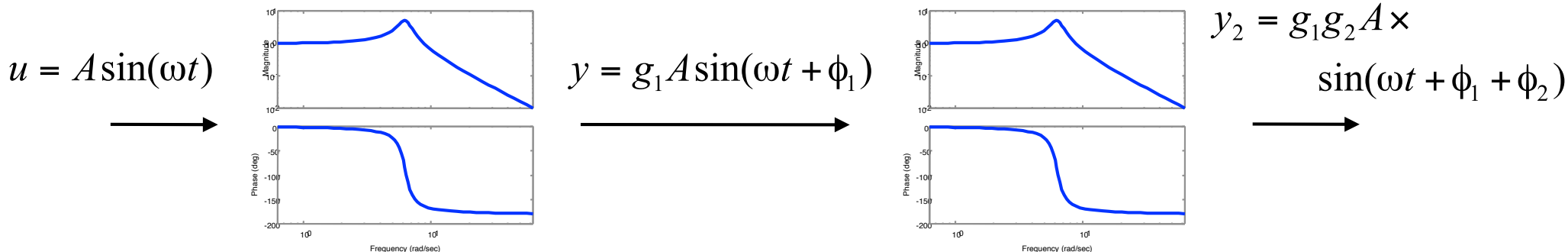
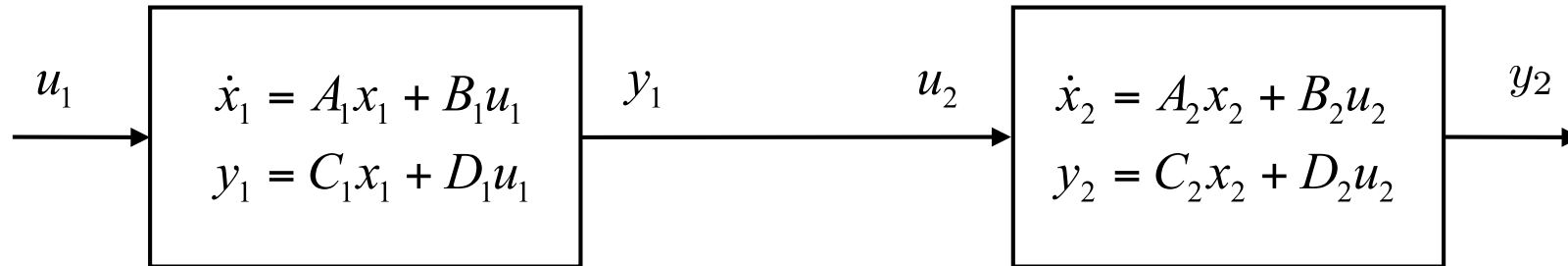
$$y = |G(i\omega)|A \sin(\omega t + \arg G(i\omega)) + \text{transients}$$

## Remarks

- Formally, can show that  $G(s)$  is the Laplace transform of the impulse response of  $\Sigma$
- Often write “ $y = G(s) u$ ” for  $Y(s) = G(s) U(s)$ , where  $Y(s)$  and  $U(s)$  are Laplace transforms of  $y(t)$  and  $u(t)$ . (Multiplication in Laplace domain corresponds to convolution.)
- Python: `G = ct.ss2tf(A, B, C, D)`

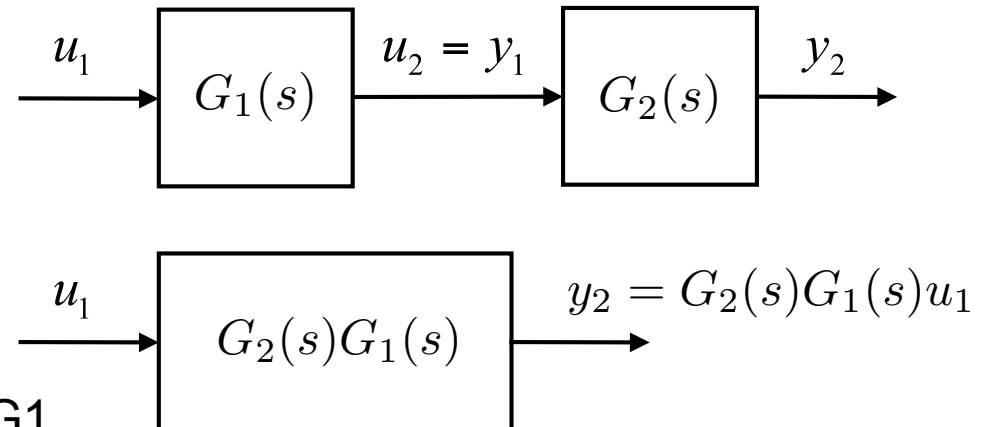
# Series Interconnections

Q: what happens when we connect two systems together in series?

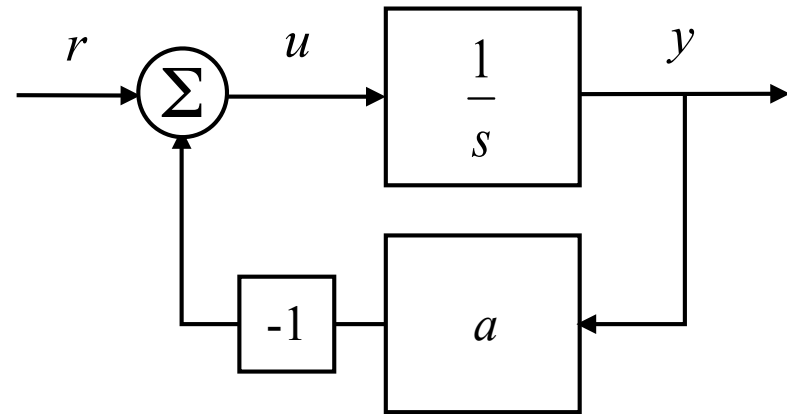
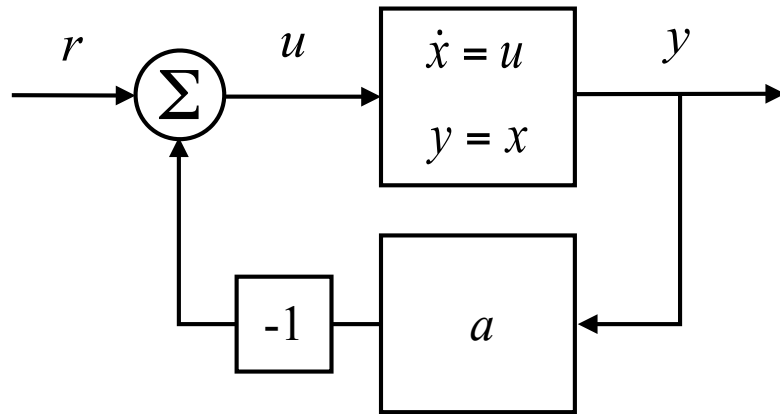


**A: Transfer functions *multiply***

- Gains multiply, phases add
- Generally: transfer functions well formulated for frequency domain interconnections
  - Convolution  $\rightarrow$  multiplication
- Python: `G = series(G1, G2)` or `G = G2 * G1`



# Feedback Interconnection



## State space derivation

$$\dot{x} = u = r - ay = -ax + r$$

$$y = x$$

## Transfer function derivation

$$y = \frac{u}{s} = \frac{r - ay}{s}$$

$$y = \frac{r}{s + a} = G(s)r$$

## Frequency response $r = A \sin(\omega t)$

$$y = \left| \frac{1}{\sqrt{a^2 + \omega^2}} \right| \sin \left( \omega t - \tan^{-1} \left( \frac{\omega}{a} \right) \right)$$

## Frequency response

$$y = |G(i\omega)| \sin(\omega t + \angle G(i\omega))$$

- Python: `G = ct.feedback(sys, a)` ← works for either state space or transfer functions

# Poles and Zeros

$$\begin{aligned}\dot{x} &= Ax + Bu \\ y &= Cx + Du\end{aligned}$$

$$G(s) = \frac{n(s)}{d(s)}$$
$$d(s) = \det(sI - A)$$

- Roots of  $d(s)$  are called *poles* of  $G(s)$
- Roots of  $n(s)$  are called *zeros* of  $G(s)$

## Poles of $G(s)$ determine the stability of the (closed loop) system

- Denominator of transfer function = characteristic polynomial of state space system
- Provides easy method for computing stability of systems
- Right half plane (RHP) poles ( $\text{Re} > 0$ ) correspond to unstable systems

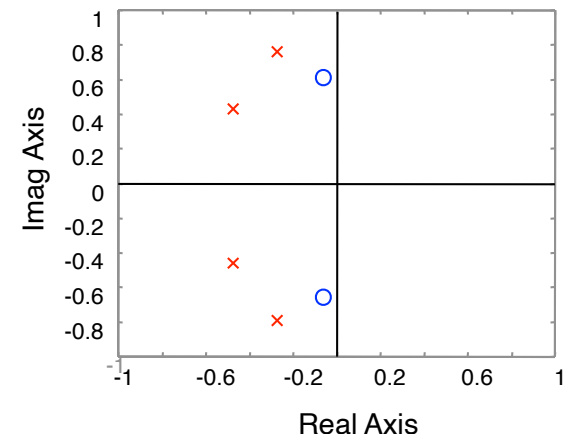
## Zeros of $G(s)$ related to frequency ranges with limited transmission

- A pure imaginary zero at  $s = i\omega$  blocks any output at that frequency ( $G(i\omega) = 0$ )
- Zeros provide limits on performance, especially RHP zeros (more on this later)

**Python:** `G.poles()`, `G.zeros()`, `ct.pole_zero_plot(G)`

$$G(s) = k \frac{s^2 + b_1s + b_2}{s^4 + a_1s^3 + a_2s^2 + a_3s + a_4}$$

`pole_zero_plot(G)`





# Sketching the Bode Plot for a Transfer Function (1/2)

Evaluate transfer function on imaginary axis

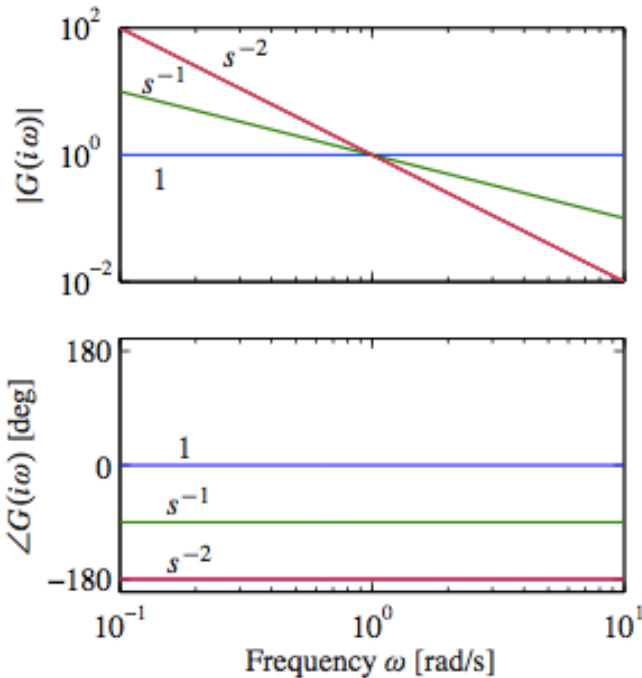
$$M = |G(i\omega)|, \quad \varphi = \arctan \frac{\text{Im } G(i\omega)}{\text{Re } G(i\omega)}$$

- Plot gain (M) on log/log scale
- Plot phase ( $\varphi$ ) on log/linear scale
- Piecewise linear approximations available

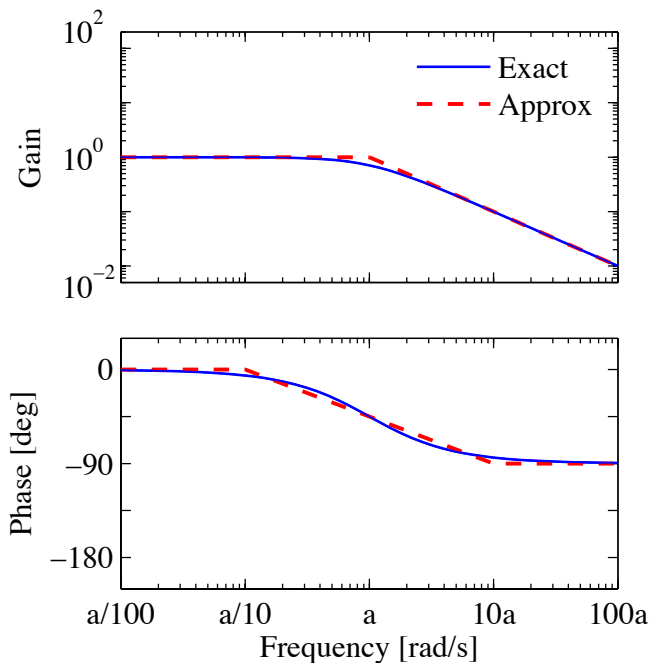
$$\log |G(i\omega)| \approx \begin{cases} 0 & \text{if } \omega < a \\ \log a - \log \omega & \text{if } \omega > a, \end{cases}$$

$$\angle G(i\omega) \approx \begin{cases} 0 & \text{if } \omega < a/10 \\ -45 - 45(\log \omega - \log a) & a/10 < \omega < 10a \\ -90 & \text{if } \omega > 10a. \end{cases}$$

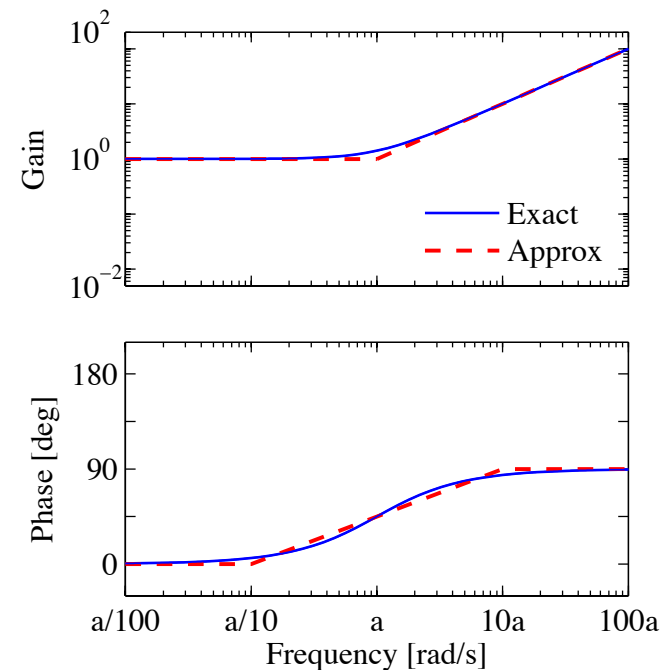
$$G(s) = \frac{1}{s^k}$$



$$G(s) = \frac{a}{s+a}$$



$$G(s) = \frac{s+a}{a}$$



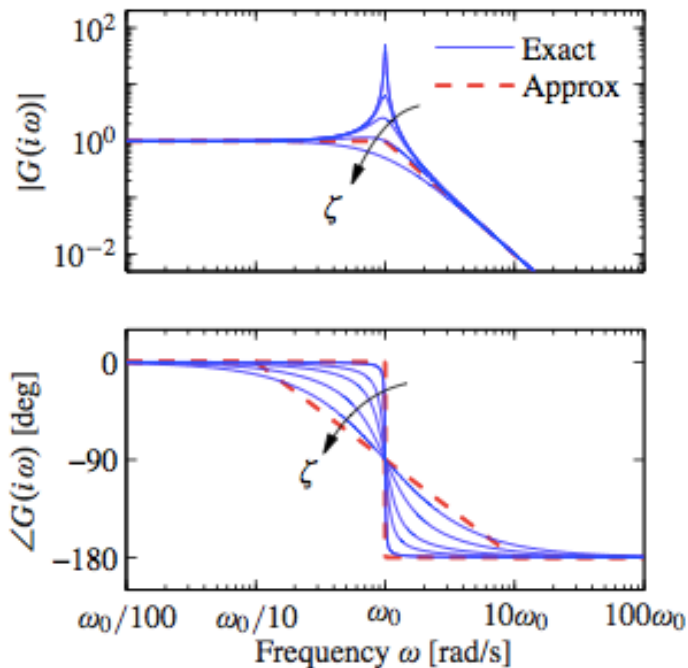
# Sketching the Bode Plot for a Transfer Function (2/2)

**Complex poles**  $G(s) = \frac{\omega_0^2}{s^2 + 2\omega_0\zeta s + \omega_0^2}$

$$\log |G(i\omega)| \approx \begin{cases} 0 & \text{if } \omega \ll \omega_0 \\ 2 \log \omega_0 - 2 \log \omega & \text{if } \omega \gg \omega_0, \end{cases}$$

$$\angle G(i\omega) \approx \begin{cases} 0 & \text{if } \omega \ll \omega_0 \\ -180 & \text{if } \omega \gg \omega_0. \end{cases}$$

$$G(s) = \frac{\omega_0^2}{s^2 + 2\omega_0\zeta s + \omega_0^2}$$

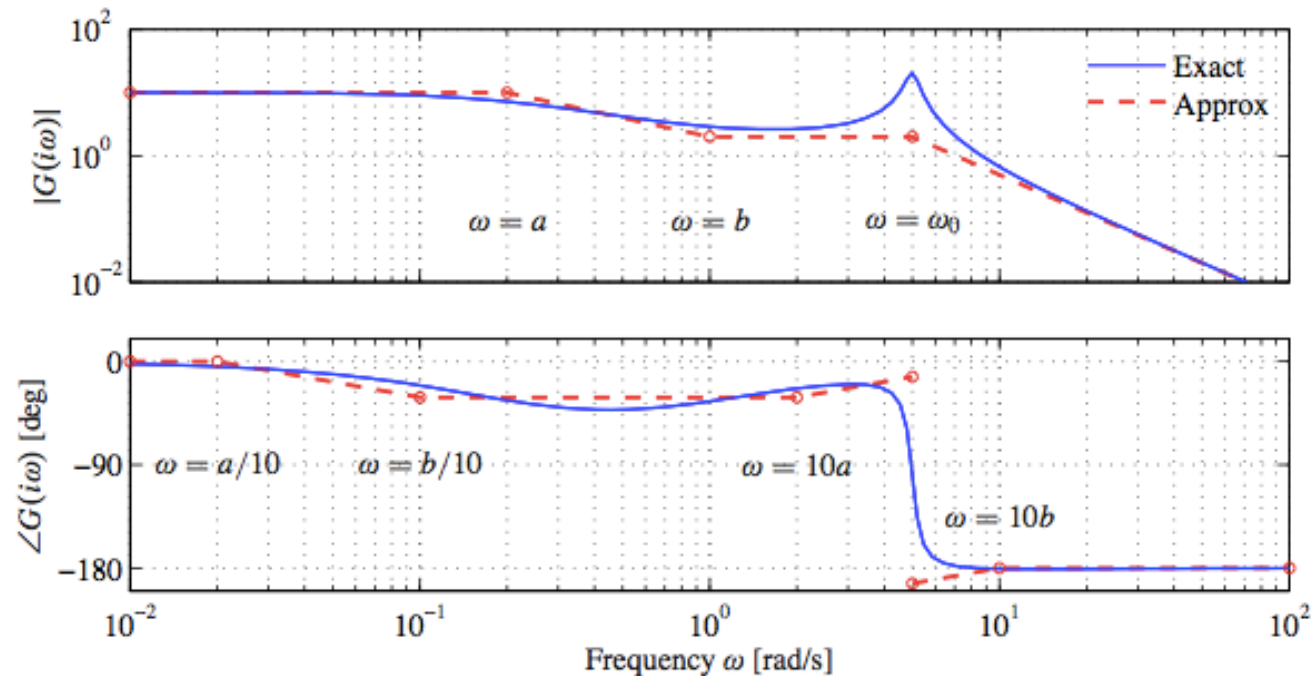


**Ratios of products**  $G(s) = \frac{b_1(s)b_2(s)}{a_1(s)a_2(s)}$

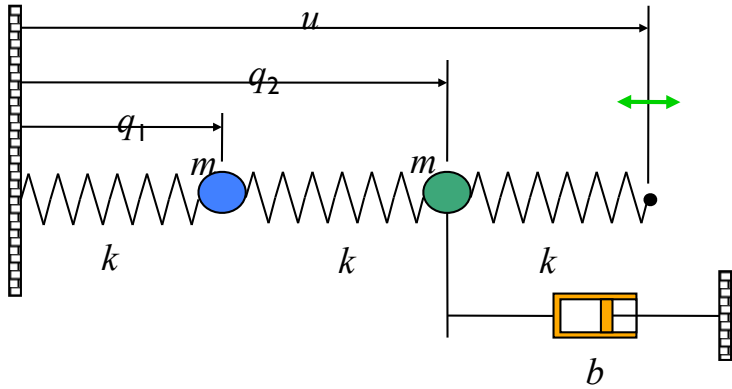
$$\log |G(s)| = \log |b_1(s)| + \log |b_2(s)| - \log |a_1(s)| - \log |a_2(s)|$$

$$\angle G(s) = \angle b_1(s) + \angle b_2(s) - \angle a_1(s) - \angle a_2(s)$$

$$G(s) = \frac{k(s+b)}{(s+a)(s^2 + 2\zeta\omega_0s + \omega_0^2)}, \quad a \ll b \ll \omega_0.$$



# Example: Coupled Masses



$$H_{q_1 f} = \frac{0.04}{s^4 + 0.08s^3 + 0.8016s^2 + 0.032s + 0.12}$$

$$H_{q_2 f} = \frac{0.2s^2 + 0.008s + 0.08}{s^4 + 0.08s^3 + 0.8016s^2 + 0.032s + 0.12}$$

## Poles ( $H_{q_1 f}$ and $H_{q_2 f}$ )

- $-0.0200 \pm 0.7743j$
- $-0.0200 \pm 0.4468j$

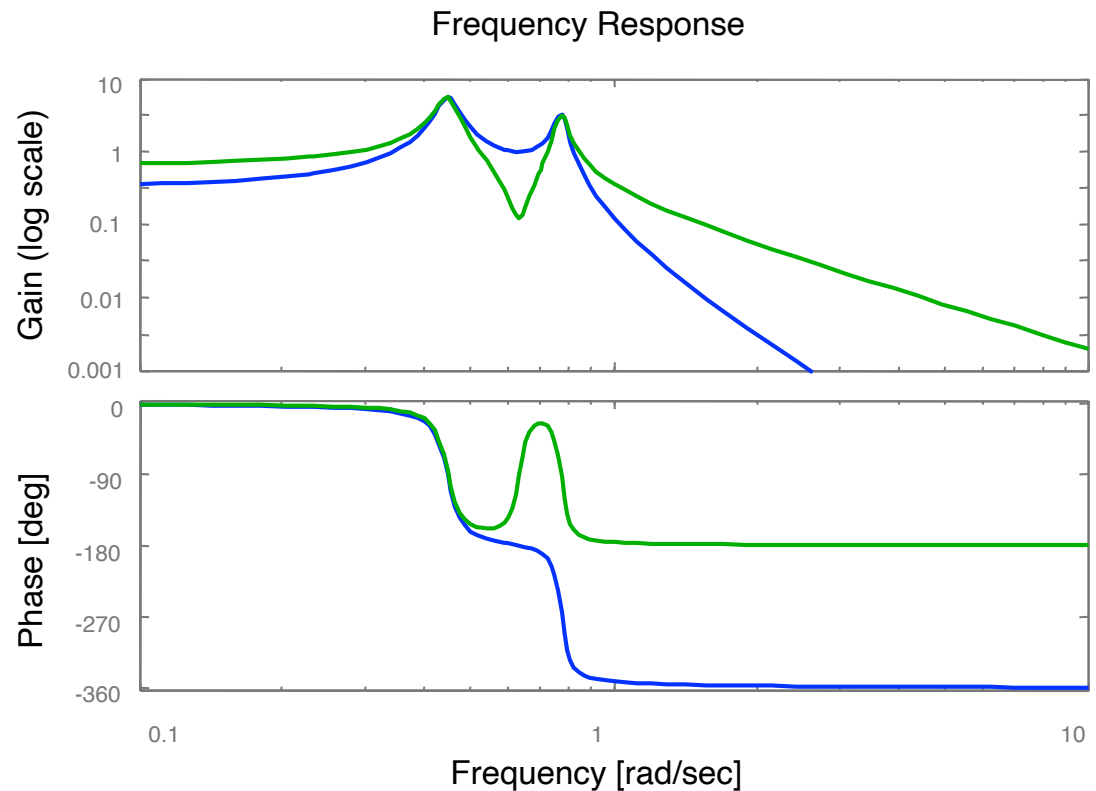
## Zeros ( $H_{q_2 f}$ )

- $-0.0200 \pm 0.6321j$

## Interpretation

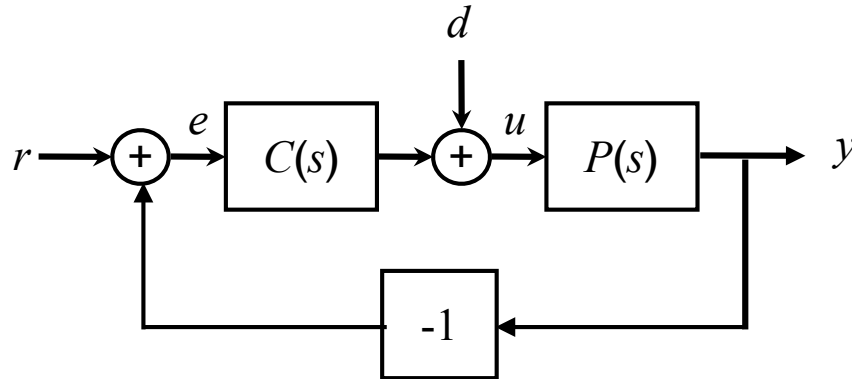
- Zeros in  $H_{q_2 f}$  give low response at  $\omega \approx 0.6321$

**Python:** `ct.bode_plot(H)` or  
`ct.frequency_response(H).plot()`



# Block Diagram Algebra

**Basic idea: treat transfer functions as multiplication, write down equations**



$$y = P(s)u$$

$$u = d + C(s)e$$

$$e = r - y$$

**Manipulate equations to compute desired signals**

$$e = r - y$$

$$= r - P(s)u$$

$$= r - P(s)(d + C(s)e)$$

$$(1 + P(s)C(s))e = r - P(s)d$$

$$e = \underbrace{\frac{1}{1 + P(s)C(s)}}_{H_{er}} r - \underbrace{\frac{P(s)}{1 + P(s)C(s)}}_{H_{ed}} d$$

Note: linearity gives superposition of terms

**Algebra works because we are working in frequency domain**

- Time domain (ODE) representations are not as easy to work with
- Formally, all of this works because of Laplace transforms (see text)

# Summary: Block Diagram Algebra

Type	Diagram	Transfer function
Series		$H_{y_2u_1} = H_{y_2u_2} H_{y_1u_1} = \frac{n_1 n_2}{d_1 d_2}$
Parallel		$H_{y_3u_1} = H_{y_2u_1} + H_{y_1u_1} = \frac{n_1 d_2 + n_2 d_1}{d_1 d_2}$
Feedback		$H_{y_1r} = \frac{H_{y_1u_1}}{1 + H_{y_1u_1} H_{y_2u_2}} = \frac{n_1 d_2}{n_1 n_2 + d_1 d_2}$

- These are the basic manipulations needed; some others are possible
- Formally, could work all of this out using the original ODEs ( $\Rightarrow$  nothing really new)

# Python (and MATLAB) Manipulation of Transfer Functions

## Creating transfer functions

- $[\text{num}, \text{den}] = \text{ct.ss2tf}(A, B, C, D)$
- $\text{sys} = \text{ct.tf}(\text{num}, \text{den})$
- $\text{num}, \text{den} = [1, a, b] \rightarrow s^2 + a s + b$

## Interconnecting blocks

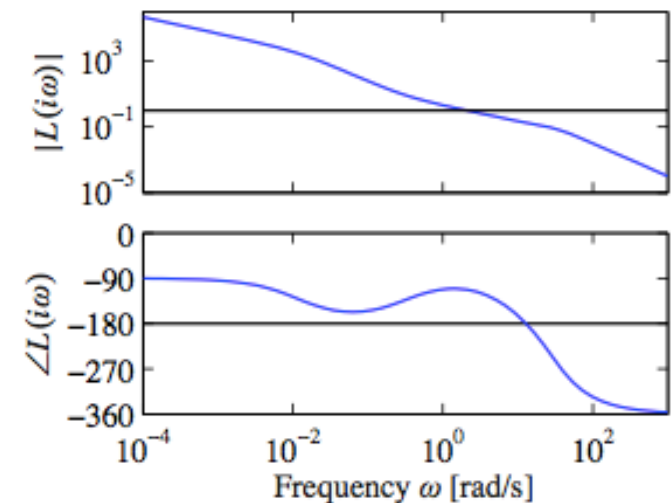
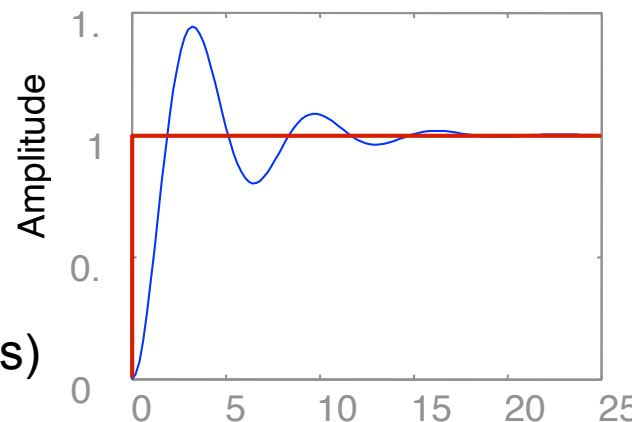
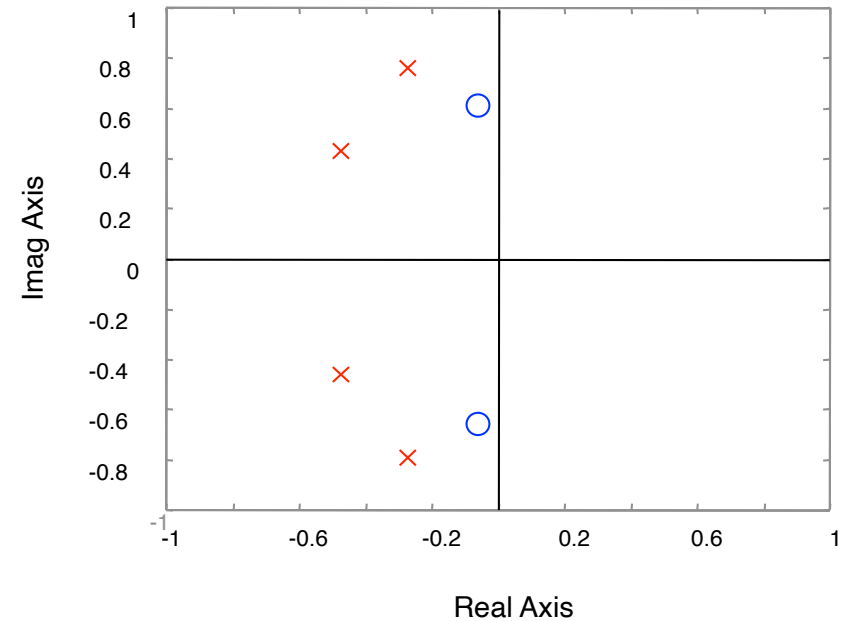
- $\text{sys} = \text{series}(\text{sys1}, \text{sys2})$ , parallel, feedback
- Also:  $\text{sys2} * \text{sys1}$  (series),  $\text{sys1} + \text{sys2}$  (par)

## Computing poles and zeros

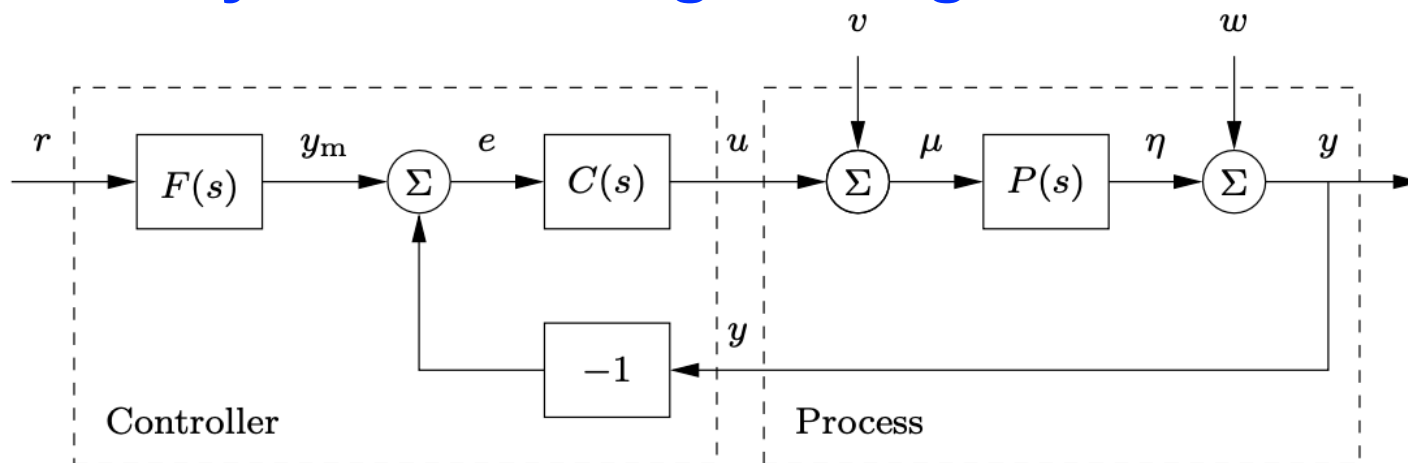
- $\text{sys.poles}()$ ,  $\text{sys.zeros}()$
- $\text{pzmap}(\text{sys})$

## I/O response

- $\text{initial\_response}(\text{sys})$
- $\text{step\_response}(\text{sys})$
- $\text{forced\_response}(\text{sys})$
- $\text{frequency\_response}(\text{sys})$



# Control Analysis and Design Using Transfer Functions



## Transfer functions provide a method for “block diagram algebra”

- Easy to compute transfer functions between various inputs and outputs
  - $H_{er}(s)$  is the transfer function between the reference and the error
  - $H_{ed}(s)$  is the transfer function between the disturbance and the error

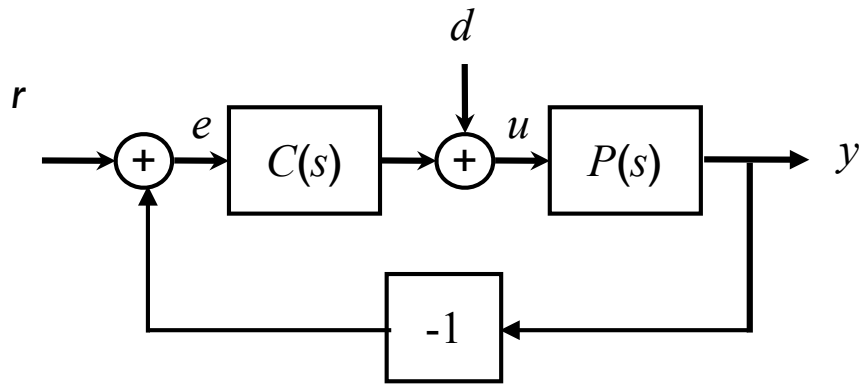
## Transfer functions provide a method for performance specification

- Since transfer functions provide frequency response directly, it is convenient to work in the “frequency domain”
  - $H_{er}(s)$  should be small in the frequency range 0 to 10 Hz (good tracking)

## Key idea: perform all analysis and design for linear systems in “frequency domain”

- Convert specifications on time response to specifications on frequency response
- “Shape” the frequency response by design of controller transfer function (Ch 10-13)

# Wednesday: Stability via Nyquist Criterion



**Determine stability from (open) loop transfer function,  $L(s) = P(s)C(s)$ .**

- Use “principle of the argument” from complex variable theory (see reading)

**Thm (Nyquist).** Consider the Nyquist plot for loop transfer function  $L(s)$ . Let

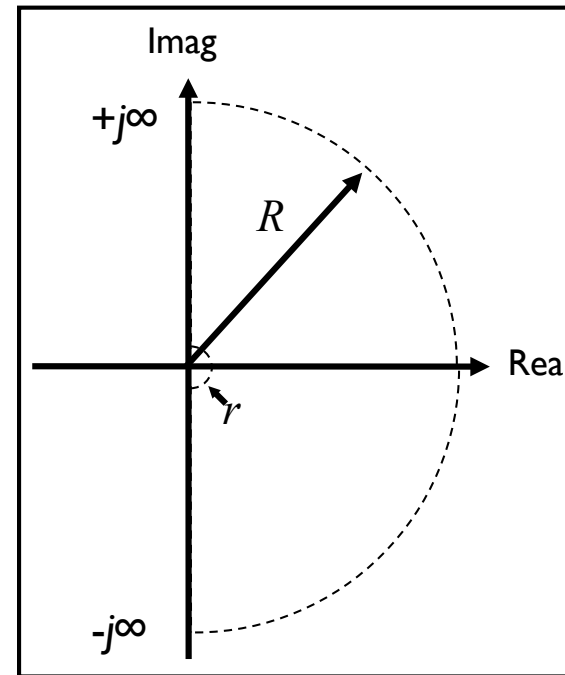
$P$  # RHP poles of  $L(s)$

$N$  # clockwise encirclements of  $-1$

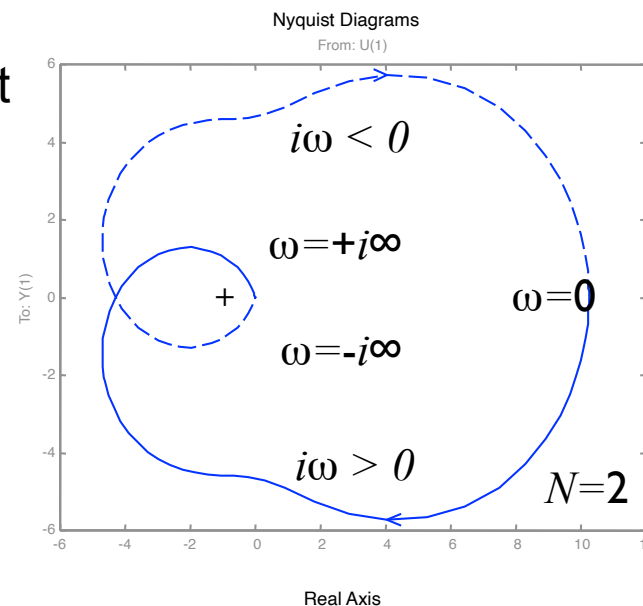
$Z$  # RHP zeros of  $1 + L(s)$

Then

$$Z = N + P$$



- Nyquist “D” contour
- Take limit as  $r \rightarrow 0, R \rightarrow \infty$
- Trace from  $-1$  to  $+1$  along imaginary axis



- Trace frequency response for  $L(s)$  along the Nyquist “D” contour
- Count net # of clockwise encirclements of the  $-1$  point