Controlling Structured Spatially Interconnected Systems

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"Recent" Developments...

Proliferation of actuators and sensors"Moore's Law"

•Embedded systems, CAN, Bluetooth

MORE PERFORMANCE!!!

What is, and will be, needed: NEW CONTROL TOOLS:

- LARGE numbers of actuators and sensors
 Distributed computation
- Limited connectivity
- Robustness
- Performance
- •Flexibility
- •Etc.

Modeling Interconnected Systems

$$\dot{x}(t) = f(x(t), d(t))$$
$$z(t) = h(x(t), d(t))$$

x(t),d(t), and z(t) live in a Hilbert space:

$$d(t) = d(t, s_1, s_2, \cdots, s_L) = d(t, \mathbf{s})$$
$$\|d(t)\|_{l_2}^2 := \sum_{s_1 = -\infty}^{\infty} \cdots \sum_{s_L = -\infty}^{\infty} d^*(t, \mathbf{s}) d(t, \mathbf{s})$$



Restrict the model class: local interconnection

WHY?

•Large class of systems, non-trivial behavior:

- vehicle platoons
- •finite difference approximations of PDEs
- •cellular automata, artificial life, etc.
- •behavior of groups, swarm intelligence, etc.

Case Study: Formation Flight

Use upwash created by neighbouring craft to provide extra lift MOTIVATION

- "satellite" type of applications (Wolfe, Chichka and Speyer '96)
- MAVs and UAVs, extend range



Lissaman and Shollenberger '70: Formation Flight of Birds

Formation Flight Test-Bed





Student: Jeff Fowler (ME)

$$\ddot{y}(t,s) = c_1 y(t,s) - c_2 \theta(t,s)$$

 $\ddot{\theta}(t,s) = -c_3 \dot{\theta}(t,s) + u(t,s) - c_4 u(t,s+1) + c_5 \dot{\theta}(t,s+1) + c_6 (y(t,s) - y(t,s+1))^2$

Define shift operator **S**:

$$(\mathbf{S}u)(t,s) \coloneqq u(t,s+1)$$

yields

$$\ddot{y} = c_1 y - c_2 \theta$$

$$\ddot{\theta} = -c_3 \dot{\theta} + u - c_4 \mathbf{S}u + c_5 \mathbf{S} \dot{\theta} + c_6 (y - \mathbf{S}y)^2$$

In general:



$$(\Delta x)(t,\mathbf{s}) = F(x(t,\mathbf{s}), d(t,\mathbf{s}), \mathbf{s})$$
$$z(t,\mathbf{s}) = H(x(t,\mathbf{s}), d(t,\mathbf{s}), \mathbf{s})$$

d(t,**s**), x(t, **s**), z(t, **s**) are FD, F and H NL functions on FD space

Special cases...

 $(\Delta x)(t,\mathbf{s}) = F(x(t,\mathbf{s}), d(t,\mathbf{s}), \mathbf{s})$ $z(t,\mathbf{s}) = H(x(t,\mathbf{s}), d(t,\mathbf{s}), \mathbf{s})$



 $F(\cdot) = Ax + Bd, H(\cdot) = Cx + Dd$ Linear, spatial invariant systems



Family of completely decentralized systems

"CONVENIENT" FRAMEWORK FOR CAPTURING STRUCTURE

Recent Related Work

Siljak et al: Decentralized control of complex systems

Bamieh, Paganini, Dahleh: Spatially Invariant Systems

Cheng, Yang, Zhai, Peterson, Savkin, ...: Decentralized Control of IC systems.

Stewart, Gorinevski, Dumont: Cross directional control

Control Design and Analysis: Spatially Invariant Systems



Stable:
$$(\Delta - A)^{-1}$$
 exists and is bounded
Contractive: $\|D + C(\Delta - A)^{-1}B\| < 1$

Analysis

 $S_1^{-1}I$

 \therefore $X_s =$

S₁*I*

 $\begin{bmatrix} X_1^{++} & X_1^{+-} \\ X_1^{-+} & X_1^{--} \end{bmatrix}$

 $\mathrm{X}^{++}_\mathrm{L}$

$$x_{\rm T} = A_{TT}x_{\rm T} + A_{TS}x_{\rm S} + B_Td$$
$$\Delta_{\rm S}x_{\rm S} = A_{ST}x_{\rm T} + A_{SS}x_{\rm S} + B_Sd$$
$$z = C_Tx_{\rm T} + C_Sx_{\rm S} + Dd$$



$$\begin{bmatrix} I & 0 & 0 \\ A_{ST}^{-} & A_{SS}^{-} & B_{S}^{-} \\ 0 & 0 & I \end{bmatrix}^{*} \begin{bmatrix} A_{TT}^{*} X_{T} + X_{T} A_{TT} & X_{T} A_{TS}^{+} & X_{T} B_{T} \\ (X_{T} A_{TS}^{+})^{*} & -X_{S} & 0 \\ (X_{T} B_{T}^{+})^{*} & 0 & -I \end{bmatrix} \begin{bmatrix} I & 0 & 0 \\ A_{ST}^{-} & A_{SS}^{-} & B_{S}^{-} \\ 0 & 0 & I \end{bmatrix} + \begin{bmatrix} I & 0 & 0 \\ A_{ST}^{+} & A_{SS}^{+} & B_{S}^{+} \\ C_{T} & C_{S} & D \end{bmatrix}^{*} \begin{bmatrix} 0 & X_{T} A_{TS}^{-} & 0 \\ (X_{T} A_{TS}^{-})^{*} & X_{S} & 0 \\ 0 & 0 & I \end{bmatrix} \begin{bmatrix} I & 0 & 0 \\ A_{ST}^{+} & A_{SS}^{+} & B_{S}^{+} \\ C_{T} & C_{S} & D \end{bmatrix}^{*} \begin{bmatrix} 0 & X_{T} A_{TS}^{-} & 0 \\ (X_{T} A_{TS}^{-})^{*} & X_{S} & 0 \\ 0 & 0 & I \end{bmatrix} \begin{bmatrix} I & 0 & 0 \\ A_{ST}^{+} & A_{SS}^{+} & B_{S}^{+} \\ C_{T} & C_{S} & D \end{bmatrix} < 0$$

Theorem: There exists a controller such that the analysis LMI is satisfied if and only if there exists structured Y and X such that

$$U^{*} \begin{bmatrix} A\mathbf{Y} + \mathbf{Y}A^{*} & \mathbf{Y}C_{1}^{*} & B_{1} \\ C_{1}\mathbf{Y} & -I & D_{11} \\ B_{1}^{*} & D_{11}^{*} & -I \end{bmatrix} U < 0$$
$$V^{*} \begin{bmatrix} A^{*}\mathbf{X} + \mathbf{X}A & \mathbf{X}B_{1} & C_{1}^{*} \\ B_{1}^{*}\mathbf{X} & -I & D_{11}^{*} \\ C_{1} & D_{11} & -I \end{bmatrix} V < 0$$

$$\begin{bmatrix} \mathbf{X}_{0} & \mathbf{I} \\ \mathbf{I} & \mathbf{Y}_{0} \end{bmatrix} > \mathbf{0}$$

Controller implementation:

$$\dot{x}_{T} = A_{TT}x_{T} + A_{TS}x_{S} + B_{T}y$$

$$\Delta_{S}x_{S} = A_{ST}x_{T} + A_{SS}x_{S} + B_{S}y$$

$$u = C_{T}x_{T} + C_{S}x_{S} + Dy$$
EX: 2D
$$x_{1}(s_{1},s_{2}) \xrightarrow{i} x_{1}(s_{1}-1,s_{2}) \xrightarrow{i} x_{S} = (x_{1}, x_{-1}, x_{2}, x_{-2})$$

$$x_{2}(s_{1},s_{2}) \xrightarrow{i} x_{3}(s_{1}-1,s_{2}) \xrightarrow{i} x_{2}(s_{1},s_{2}) \xrightarrow{i} x_{3}(s_{1}-1,s_{2}) \xrightarrow{i} x_{3}(s_{1}-1,s_{2}) \xrightarrow{i} x_{3}(s_{1}-1,s_{2}) \xrightarrow{i} x_{3}(s_{1}-1,s_{2}) \xrightarrow{i} x_{4}(s_{1},s_{2}) \xrightarrow{i} x_{4}(s_{1},s_{2}) \xrightarrow{i} x_{4}(s_{1},s_{2}) \xrightarrow{i} y(s_{1},s_{2})$$

Control Architecture



Decentralized Control



Decentralized Control

Distributed Control



COMPARISON OF SPATIAL 2-NORM, ROLL ANGLE



STRONG NONLINEAR COUPLING



Nonlinear Spatially Interconnected Systems:

$$\Delta x = f(x) + g(x)d$$
$$z = h(x)$$

Feedback linearizationBacksteppingetc.

Other ongoing work Spatially and Time Varying Systems:

non-homogeneous propertiesfinite boundary conditions

$$\Delta x = A(s)x + B(s)d$$
$$z = C(s)x + D(s)d$$

TOOLS:

- LTI to LTV machinery (GEIR DULLERUD, UIUC)
- method of images, etc.
- LPV tools



Framework for Robust Control of IC systems

$$(\Delta x)(t,\mathbf{s}) = F(x(t,\mathbf{s}), d(t,\mathbf{s}), \mathbf{s})$$
$$z(t,\mathbf{s}) = H(x(t,\mathbf{s}), d(t,\mathbf{s}), \mathbf{s})$$

Delta contains temporal operators, spatial operators, AND uncertainty.

Student: Ramu Chandra (AE)

Model Reduction (CAROLYN BECK, UIUC)

Cross-Directional Control (GREG STEWART, HONEYWELL)

Phased Array Antennas for AFV Communication Student: Sean Breheny (ECE)

High data rate comms between AFVs and base station/satellite (video, etc.)

- Difficult to put a high gain antenna on an AFV (size constraint)
- Since it may be advantageous to use groups of AFVs anyway, why not investigate whether a formation of AFVs, each carrying a low gain antenna, could form a high gain phased array?



What is a Phased Array Antenna?

Exploit EM wave interference among several antennas.

- For the simplest case (where array elements are not strongly coupled to each other), gain increases roughly linearly in N, the number of elements.
- Channel capacity increases linearly when the maximum bandwidth is used.



Example: Endfire Array

10 Element Endfire Array, Nominal Gain=8.1 Red - Original, Blue - Uncorrected, Black - Corrected NOTE: Radial axis is linear



